

REMARKS

Prior to this Reply, Claims 1-19 and 21-42 were pending. Through this Reply, Claims 1, 5-16, 18, 19, 21, 25-33, 41 and 42 have been amended; Claims 2-4, 17, 22-24 and 34-40 have been cancelled without prejudice to, or disclaimer of, the subject matter contained therein; and, Claims 43-85 have been added. Accordingly, Claims 1, 5-16, 18, 19, 21, 25-33 and 41-85 are now at issue in the present case.

I. Claims Rejections

The Examiner rejected Claim 38 under 35 U.S.C. 102(e) as being anticipated by U.S. Patent No. 6,288,856 to Ottesen et al. (hereinafter “Ottesen”).

The Examiner also rejected Claims 1, 2, 5, 7, 16-19, 21, 22, 26 and 37 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen in view of U.S. Patent No. 4,929,894 to Monett (hereinafter “Monett”).

The Examiner also rejected Claim 39 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen in view of U.S. Patent No. 6,100,683 to Lim (hereinafter “Lim”).

The Examiner also rejected Claims 3, 4, 23, 24 and 25 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen and Monett in view of Lim.

The Examiner also rejected Claims 6, 8, 25 and 32 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen and Monett in view of U.S. Patent No. 6,252,731 to Sloan (hereinafter “Sloan”).

The Examiner also rejected Claims 9 and 27 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen and Monett in view of U.S. Patent No. 6,513,141 to Livingston (hereinafter “Livingston”).

The Examiner also rejected Claims 10, 12, 28 and 30 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen, Monett and Livingston in view of Sloan.

The Examiner also rejected Claims 13, 14, 31, 33, 34 and 36 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen and Monett in view of U.S. Patent No. 5,793,548 to Zook (hereinafter “Zook”).

The Examiner also rejected Claim 15 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen, Monett and Zook in view of U.S. Patent No. 6,381,203 to Muramatsu (hereinafter “Muramatsu”).

The Examiner also rejected Claim 39 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen in view of U.S. Patent No. 6,208,476 to Park (hereinafter “Park”).

The Examiner also rejected Claim 40 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen in view of Zook.

The Examiner also rejected Claim 41 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen in view of Muramatsu.

The Examiner also rejected Claim 42 under 35 U.S.C. § 103(a) as being unpatentable over Ottesen in view of U.S. Patent No. 6,411,452 to Cloke (hereinafter “Cloke”).

Ottesen discloses a disk drive that estimates head-to-disk clearance (fly height) to identify poorly performing heads. The disk drive exploits the principle that narrower pulses represent smaller head-to-disk clearance and wider pulses represent larger head-to-disk clearance. The disk drive calculates the differentiated edge slope, the area portion, or the width of several isolated pulses to determine the pulse shapes, and then averages these calculations to reduce variations due to disk coersivity and air turbulence. Furthermore, the isolated pulses are substantially free of intersymbol interference and filter dynamics to avoid shape alterations.

Monett discloses a disk drive that simultaneously performs missing pulse and additional pulse tests on a disk to increase quality control throughput. An input waveform is delayed by one period, and a difference between the input waveform and the delayed version is compared with a missing pulse threshold and an additional pulse threshold.

Claims 1 and 21 have been amended to further clarify the claimed invention.

Claim 1 recites “deriving a value from m of the n samples, wherein the m samples are significant samples that each have an amplitude greater than 50% of an amplitude of an isolated pulse in the read signal and greater than an amplitude of the other samples of the n samples; comparing the derived value to a threshold value; and reporting a flaw in the track if the comparison is unacceptable.” Claim 21 recites similar limitations, and the remaining rejected claims that are pending depend from Claims 1 or 21.

Ottesen fails to teach or suggest detecting a flaw in a disk, and Monett fails to teach or sampling a read signal.

Ottesen in view of Monett might suggest a disk drive that estimates head-to-disk clearance to identify poorly performing heads (in Ottesen) and simultaneously performing missing pulse and additional pulse tests to rapidly identify flaws in the disk (in Monett). However, the proposed combination fails to teach or suggest sampling a read signal to detect flaws in the disk, much less using the significant samples set forth in Claim 1.

The Examiner admits that “Ottesen et al. does not explicitly teach [using the isolated pulses to detect a flaw in the disk].” However, the Examiner asserts that “It would have been obvious to a person of ordinary skill in the art, at the time the invention was made, to modify Ottesen et al.’s invention with the teaching of Monett in order to detect flaws in the disk by the examination of the track data.”

Ottesen is non-analogous to the present invention. Ottesen is directed to estimating head-to-disk clearance to identify poorly performing heads. The present invention, on the other hand, is directed to detecting flaws in a disk using significant samples of a read signal. Ottesen is neither within the field of the invention (detecting flaws in a disk) nor pertinent to the particular problem with which the inventor was concerned (increasing the signal-to-noise ratio in the samples compared to the threshold value to more accurately distinguish between flaws and noise). Therefore, Ottesen is non-analogous to the present invention and cannot be used to sustain an obviousness rejection.

Ottesen fails to teach or suggest “deriving a value from m of the n samples, wherein the m samples are significant samples that each have an amplitude greater than . . . an amplitude of the other samples of the n samples.” Instead, Ottesen estimates the head-to-disk clearance using all of the incoming isolated pulses and makes no attempt to select isolated pulses with larger amplitudes than other isolated pulses to calculate the estimate. Furthermore, selecting the isolated pulses with the largest amplitudes is senseless since the isolated pulses have identical amplitudes (see Fig. 5) and are substantially free of intersymbol interference and filter dynamics to avoid shape alterations.

Ottesen fails to teach or suggest using the isolated pulses to detect disk flaws, and Monett fails to cure this deficiency. Ottesen estimates the head-to-disk clearance based on the principle that narrower pulses represent smaller head-to-disk clearance and wider pulses represent larger head-to-disk clearance (see Fig. 6). However, neither Ottesen nor Monett even remotely suggest that the pulse shape is indicative of disk flaws. Thus, the proposed modification in which Ottesen detects disk flaws like head-to-disk clearance is completely unsupported and strikes Applicant as classic hindsight reconstruction using the present invention as a blueprint.

Ottesen teaches away from using the head-to-disk clearance estimate to detect disk flaws:

Due to minute local coercivity variations in the magnetic surface coating around a track of interest, such as a track containing di-bits having appropriate properties, and small changes in head-to-disk spacing due to air turbulence, there may be some variation in the maximum value of the derivative signal, $y[n]$. It is therefore desirable to average the maximum values from several isolated di-bits on one or more tracks. (Col. 9, lines 19-26).

As set forth in MPEP § 2141, in order to be analogous art, the reference must either be in the field of the invention or be reasonably pertinent to the particular problem with which the inventor was concerned.

As set forth in MPEP § 2143, in order to establish a *prima facie* case of obviousness, three basic criteria must be met. First, there must be some suggestion or motivation, either in the references themselves or in the knowledge generally available to one of ordinary skill in the art, to modify the reference or to combine reference teachings. Second, there must be a reasonable expectation of success. Finally, the prior art reference (or references when combined) must teach or suggest all of the claim limitations.

As set forth in MPEP § 2143, if the proposed modification would render the prior art unsatisfactory for its intended purpose, then there is no suggestion or motivation to make the proposed modification.

As set forth in MPEP § 2142, the Examiner bears the initial burden of factually supporting any *prima facie* conclusion of obviousness.

In view of the above, Applicant believes that the Examiner has failed to meet this burden. Ottesen is non-analogous to the present invention, fails to teach or suggest numerous features of the present invention, and teaches away from a proposed modification that has no support in Monett or any other cited reference and is probably inoperative.

Many of the dependent claims are patentable for reasons in addition to those provided above. Applicant does not admit that the Examiner has met the burden of establishing a *prima facie* case of obviousness with respect to the dependent claims. Accordingly, Applicant may address the dependent claims at a later time, if necessary.

II. New Claims

Claims 43-85 have been added. No new matter has been added.

Claims 43-45 depend from Claim 21 and are allowable for at least the same reasons as Claim 21.

Claim 46 is allowable for at least the same reasons as Claim 1. Claims 47-65 depend from Claim 46 and are allowable for at least the same reasons as Claim 46.

Claim 66 is allowable for at least the same reasons as claim 46. Claims 67-75 depend from Claim 66 and are allowable for at least the same reasons as Claim 66.

Claim 76 is allowable for at least the same reasons as Claim 46. Claims 77-85 depend from Claim 76 and are allowable for at least the same reasons as Claim 76.

III. Other Amendments to Claims

The claims have been amended to improve clarity. No new matter has been added.

IV. Amendments to Specification

A substitute specification without claims (and a marked-up version thereof) is provided herein under 37 C.F.R. 1.125 to improve clarity of the specification. No new matter has been added.

Applicants respectfully request that the substitute specification be entered.

V. Amendments to Drawings

Applicants are submitting replacement Figures 1-3, 4B, 4C and 5-7 (on Replacement Sheets 1-5) to improve the quality of the drawings.

Figure 2 has been modified to delete reference numeral 136a.

Figure 3 has been modified at step 300 to change “polarizations” to “data pattern” and at step 304 to change “take n-1” to “read data pattern from disk to obtain n-1” and at step 308 to change “take a” to “read data pattern from disk to obtain” and at step 312 to change “last” to “previous” and to insert “value” after “threshold” and to delete “are” and “the” and at step 316 to change “signal controller” to “report flaw” and to insert a branch from step 316 to step 308.

Figure 4B has been modified so that the magnetic transitions between magnets 404 are aligned with the magnetic transitions between bit cells 400 in Figure 4A.

Figure 4C has been modified so that peaks 412 are aligned with the magnetic transitions between magnets 404 in Figure 4B and the magnetic transitions between bit cells 400 in Figure 4A.

Figure 5 has been modified to change “A to D sample levels (no Noise)” to “Amplitude”.

Figure 6 has been replaced by a new Figure 6.

Figure 7 has been modified at shift register 700 to change “n” to “m” and at comparator 708 to insert “value” after “threshold” and at memory 712 to change “valve” to “value”.

No new matter has been added. Figures 1-3, 4A, 4B, 4C and 5-7 constitute all of the drawings of the application.

VI. Additional Claim Fees

In determining whether additional claim fees are due, reference is made to the Fee Calculation Table (below).

Fee Calculation Table						
	Claims Remaining After Amendment		Highest Number Previously Paid For	Present Extra	Rate	Additional Fee
Total (37 CFR 1.16(c))	70	Minus	42	= 28	x \$50 =	\$ 1400.00
Independent (37 CFR 1.16(b))	5	Minus	3	= 2	x \$200 =	\$ 400.00

As set forth in the Fee Calculation Table (above), Applicant previously paid claim fees for forty-two (42) total claims and for three (3) independent claims. Therefore, Applicant hereby authorizes the Commissioner to charge the credit card identified on the enclosed Form PTO-2038 in the amount of \$1800.00 for the presentation of twenty-eight (28) total claims over forty-two (42) and two (2) independent claims over three (3). Although Applicant believes that no other fees are due, the Commissioner is hereby authorized to charge Deposit Account No. 50-2198 for any fee deficiencies associated with filing this paper.

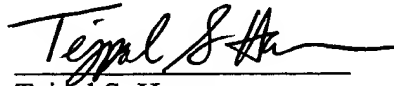
VII. Conclusion

It is believed the above comments establish patentability. Applicant does not necessarily accede to the assertions and statements in the Office Action, whether or not expressly addressed.

Applicant believes that the application appears to be in form for allowance. Accordingly, reconsideration and allowance thereof is respectfully requested.

The Examiner is invited to contact the undersigned at the below-listed telephone number regarding any matters relating to the present application.

Respectfully submitted,

A handwritten signature in black ink, appearing to read "Tejpal S. Hansra", written over a horizontal line.

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Marked-Up Version of

Substitute Specification Under 37 C.F.R. 1.125

5 **METHOD AND APPARATUS FOR FLAW DETECTION IN DISK DRIVE**
 SYNCHRONOUS SAMPLING (PRML) READ CHANNELS USING
 SIGNIFICANT SAMPLES OF DATA PATTERN STORED ON DISK
 POST PROCESSED DIGITAL FILTERS

CROSS-REFERENCE TO RELATED APPLICATIONS

10 This application claims p~~Priority is claimed~~ from U.S. Provisional Patent
Application Serial No. 60/203,088, filed May 9, 2000, ~~entitled "ENHANCED FLAW~~
~~DETECTION IN SYNCHRONOUS SAMPLING (PRML) READ CHANNELS USING~~
~~POST PROCESSED DIGITAL FILTERING"~~ and further identified as Attorney Docket
No. 3123-354 PROV, the disclosure of which is incorporated herein by reference in its
15 entirety.

FIELD OF THE INVENTION

 The present invention relates to flaw ~~the detection of flaws in~~ storage ~~magnetic~~
media, and i ~~using post processed digital filtering.~~ In particular, ~~the present invention~~
20 ~~relates to~~ flaw ~~the detection of flaws in a disk in a disk drive magnetic media using~~
samples generated by reading a data pattern on the disk ~~in a synchronous sampling,~~
~~partial response maximum likelihood (PRML) read channel using post processed digital~~
filtering.

25 BACKGROUND OF THE INVENTION

~~Computer~~ disk drives store information on magnetic disks. Typically, the information is stored ~~on each disk~~ in concentric tracks on the disk and the tracks that are divided into servo sectors that store servo information and data fieldsectors that store user data. A transducer head reads from and writes to ~~Information is written to or read from~~ thea disk. The t ~~by a transducer head is~~ mounted on an actuator arm assembly that ~~capable of moving~~ the transducer head radially over the disk. Accordingly, the ~~movement of the~~ actuator arm assembly allows the transducer head to access different tracks on the disk. The ~~A~~ disk is rotated by a spindle motor at a high speed, allowing the transducer head to access different data fieldsectors within each track on the disk. ~~The~~ transducer head may include integrated read and write heads.

Fig. 1 illustrates a ~~A typical computer disk drive 100 that is illustrated in Fig. 1.~~ The ~~disk drive 100~~ includes a base 104 and a magnetic disk (or disks) 108 (only one of which is shown ~~in Fig. 1~~). The ~~magnetic disks 108~~ is ~~are~~ interconnected to the base 104 by a spindle motor (not shown) mounted within or beneath ~~at~~ the hub 112 such that the disks 108 ~~can be rotated~~ relative to the base 104. An ~~actuator arm assemblies 116~~ is ~~(only one of which is shown in Fig. 1)~~ are interconnected to the base 104 by a bearing 120 and suspends. ~~Actuator arm assemblies 116~~ each include a transducer head 124 at a first end. The transducer head 124 reads data from and writes data to, ~~to address each of the surfaces of the magnetic disks 108. The transducer heads 124 typically include read and write elements (not shown).~~ A voice coil motor 128 pivots the actuator arm assemblies 116 about the bearing 120 to radially position the transducer heads 124 relativewith respect to the ~~magnetic~~ disks 108. By changing the radial position of the transducer heads 124 relativewith respect to the ~~magnetic~~ disks 108, the transducer heads

124 can access different tracks or cylinders 132 on the magnetic disks 108. The voice coil motor 128 is operated by a controller 136 that is in turn operatively connected to a host computer (not shown). A channel 140 processes data information read from the magnetic disks 108 by the transducer heads 124.

5 ~~With reference now to Fig. 2, illustrates the a typical arrangement of data tracks~~
~~132 on a magnetic disk 108 in more detail is illustrated. Usually, the data tracks 132 are~~
~~divided into data fields 204a-204h and with a servo sectors 208a-208h between one or~~
~~more of the data fields 204a-204h. Generally, the data fields 204a-204h are used for~~
~~storing user data and, while the servo sectors 208a-208h are used for storing servo~~
10 ~~information, that is used to provide the transducer head 124 with its radial positioning~~
~~information. Typically, at least some of the information contained in the servo sectors~~
~~208a-208h is written during the servo track writing process, and the portions of the servo~~
~~sectors 208a-208h containing such information generally cannot be written to after the~~
~~disk drive 100 is assembled. In particular, the servo sectors 208a-208h provide the~~
15 ~~transducer heads 124 with information concerning their position over the magnetic disks~~
~~108.~~

Although the magnetic disk 108 illustrated in Figs. 1 and 2 is shown as having a relatively small number of data tracks 132, data fields 204 and servo sectors 208, it can be appreciated that a typical computer disk drive 100 contains a very large number of
20 data tracks 132, data fields 204 and servo or hard sectors 208. For example, computer disk drives 100 having over 30,000 tracks per inch and 120 servo sectors per track are presently available.

———In addition, alternate configurations of the magnetic disks 108 are possible. For example, ~~in a computer disk drive 100 having several magnetic disks 108, one~~ a surface of ~~one of the disks 108 can~~ may be dedicated to servo information; while the other surfaces of the ~~remaining disks 108 (and any remaining disks 108 in the disk drive 100) can~~ may be used exclusively to store user data.

Data is ~~generally stored on the disk 108 using data patterns of magnetization patterns with magnetic transitions between opposite magnetic polarities.~~ For example, the magnetic polarization of the disk 108 in a first direction may encode a digital 1, and while the magnetic polarization in a second direction may encode a digital 0. A bit cell is the shortest length of the track 132 to which a particular magnetic polarity is written. Accordingly, a ~~transition in magnetization~~ transition from one bit cell to ~~the~~ next bit cell indicates signals a change from one digital character to another. ~~A bit cell is the shortest length of a track 132 to which a particular magnetic polarity is written.~~

15 The magnetic disk 108 ~~is~~ is generally formed ~~from a film of magnetically hard material deposited on a substrate.~~ For example, ~~the disk 108 may be formed by~~ depositing a magnetic metal film on a rigid substrate. The thickness of the magnetic film must be closely controlled. Where, ~~for example,~~ the magnetic film is too thin, the magnetic flux density produced by a magnetic transition ~~in the thinned area~~ will be too weak less than the magnetic flux produced by a magnetic transition in an area of the disk ~~having the specified film thickness.~~ The disk 108 may also contain other defects, such as scratches or pits, that degrade the magnetic flux density produced by of a magnetic transition. These defects can ~~may also be affected by other defects on the disk 108, such~~

as scratches or pitting occurring during the manufacture of the disk 108, or during the assembly of the disk drive 100.

~~In order to ensure the reliable storage and retrieval of user data from a disk drive 100, manufacturers generally subject a~~ The disk drive 100 is subject to numerous qualification tests to ensure reliable storage and retrieval of user data once before the disk drive 100 is delivered to an end user. Flaw scan is one such qualification test. Flaw scan testing is generally conducted in order to identify areas of the disk 108 that may not reliably store ~~encode~~ user data. Flaw scan writes a data pattern ~~According to conventional methods of detecting flaws, a pattern of magnetic polarizations is written to those areas of the disk 108 to which write operations are allowed (e.g., to the data fields~~ sectors 204 (and any other writable areas of the disk 108) and then reads the data pattern from the data fields 204 (and any other writable areas of the disk 108) } following assembly of the disk drive 100. The magnetic polarityzation in the data pattern can ~~may be alternated~~ every bit cell to produce in the data sectors (i.e., a 1T data pattern,); ~~or may be alternated after every ith bit cell to produce to produce an iT data pattern, where i is an integer number.~~ For instance, the magnetic polarityzation can ~~may be alternated~~ every two bit cells to produce a 2T data pattern (110011001100 . . .), every three bit cells to produce a 3T data pattern. ~~A 3T data pattern is produced if the data is written to the disk in a (~~ 111000111000 . . .) and so on pattern.

The transducer head 124 generates a read signal in response to reading the data pattern from the disk 108, and the read signal includes pulses caused by the magnetic transitions in the data pattern. The isolated pulse width (PW50) is the distance between the points of intersection between an isolated pulse and a line indicating 50% of the

maximum amplitude of the isolated pulse. Intersymbol interference is the alteration of an isolated pulse due to linear superposition of other pulses in close proximity.

Data patterns —~~In general, the higher the value of i the greater the signal amplitude, because of decreased intersymbol interference. This is true for iTs that result in data patterns having with long periods (iT) that occupy a length of the track 132 that is greater than the PW50 (*i.e.*, the isolated pulse width) of a read signal derived from the disk 108 cause the transducer head 124 to generate a read signal with greater amplitude due to decreased intersymbol interference. Alternatively, data patterns with short periods that occupy a length of the track 132 that is less than the PW50 increase the likelihood of detecting a flaw or the inability of a particular length of the track 132 to produce the prescribed magnetic flux density.~~

~~The PW50 is the distance between the points of intersection between an isolated pulse and a line indicating an amplitude that is equal to 50% of the maximum amplitude of the isolated pulse. Alternatively, a shorter period increases the likelihood of detecting a flawed area or the inability of a particular length of track 132 to produce a prescribed magnetic flux.~~

~~———A conventional method for detecting flaws in a disk 108 is illustrated in Fig. 3. According to such a method, after the chosen pattern of magnetic polarizations has been written to the writable areas of the disk 108 by the write element of the transducer head 124 (step 300), the read element of the transducer head 124 is used to read back the pattern of magnetic polarizations. In general, the channel 140 samples the varying voltage signal produced by the read element of the transducer head 124 from the pattern~~

of magnetic polarizations written to the disk 108 as the track 132 passes beneath the transducer head 124.

In a typical disk drive 100, the channel 140 includes a partial response maximum likelihood (PRML) detector, (not shown) that which allows the accurately
5 detection the of data patterns of magnetic polarizations even when the user data is
densely written on the disk 108 at high bit density and the read signal exhibits
intersymbol interference. The PRML In a partial response maximum likelihood type
detector samples, the amplitude of read a signal is sampled at regular time intervals and -
determines a A-code word that symbolizes a set of pulses symbolized by a set of pulses is
10 then determined using a statistical maximum likelihood or Viterbi process. Accordingly,
the targeted shape of a pulse is determinative in decoding a stored code word. The
advantage of the PRML type detector is that the density of data, known as the user bit
density (UBD), may be increased as compared to peak detection methods.

—In general, peak detection methods must detect discrete bits of data. Thus, when
15 pulses are densely packed such that the signal derived from a first bit is distorted, e.g., by
intersymbol interference, peak detection methods are incapable of reliably decoding
pulses. In contrast, the PRML type detection method allows for the accurate detection of
bits even when the pulse that would be generated by a bit in isolation is altered by its
proximity to other bits (i.e., is altered by intersymbol interference due to the linear
20 superposition of the pulses). For instance, the PRML detector disk drives may accurately
detects a data pattern series of bits when the isolated pulse width (PW50) of a signal
contains 2.5 bits of information. Accordingly, thea PRML detector allows user data to be
recorded at higher density than a peak detector since the peak detector is incapable of

reliably decoding pulses with intersymbol interference allows for information to be recorded at increased user bit densities.

Flaw scanning techniques typically take advantage of the characteristics of the channel. For example, he channel 140 often uses a 2T preamble is often used by the channel to synchronize sample times (*i.e.* phase) and to determine signal the amplitude of signals in the channel to adjust allow the gain to be properly adjusted. When the phase and gain are properly adjusted, a 2T sampled waveform in the channel 140 produces a very distinctive pattern. Furthermore, In a conventional flaw scan often uses ning technique, a the 2T data pattern is often used because of the high magnetic flux transition rates, the relative-low intersymbol interference, availability in the channel 140 and the unique sampled pattern it produces in the channel 140.

Fig. 3 is a flow chart of a conventional flaw scan. The transducer head 124 writes a data pattern to the data fields 204 (step 300) and then reads the data pattern from the data fields 204 to obtain ——— With reference now to Fig. 3, according to conventional

flaw detection techniques, a pattern of magnetic polarizations is written to the disk (step 300). The channel 140 may then analyze the last n samples of the signal derived by the read element of the transducer head 124 from the disk 108. In general, $n-1$ samples are taken (step 304) and t . Then, a next sample (the n^{th} sample) is taken (step 308). The channel 140 serially determines whether each of the previous n samples under

consideration have an amplitude that is less than a threshold value (step 312). If at least one of the previous last n samples has an amplitude that is greater ~~more~~ than the threshold value, then the channel 140 ~~system~~ returns to step 308 to take a next sample. Otherwise,

the channel 140 reports a flaw to the controller 136 (step 316) and returns to step 308 to take a next sample.

~~C~~Therefore, ~~this conventional technique for detecting flaw scan s~~ is very susceptible to erroneously qualifying a series of bit cells where noise or some other disturbance ~~may causes~~ one or more samples to exceed the threshold value. As a result accordingly, areas of the disk 108 that cannot reliably store user data may nonetheless be qualified ~~to store data due to the effects of noise~~. Additionally, although the disk drives 100 uses ~~are typically provided with error correctionng code (ECC) that allows the disk drive 100 to tolerate at least some errors, the storage ability of such error~~ correcting code to allow for the reliability ~~le storage of data in areas of the disk 108 containing flaws could still be compromised~~. Similarly, ~~this conventional flaw scan method~~ is susceptible to erroneously disqualifying a length of the track 132 that does not contain errors in the presence of a a-sustained noise event that causes a series of samples to fall below the threshold value. This unnecessarily reduces the storage capacity of the disk drive 100.

Conventional flaw scan ~~In order to reduce the effects of noise, conventional flaw scanning is typically makes performed by making two or more passes over each surface of every disk 108 included in the a disk drive 100 to reduce~~. ~~Multiple flaw detection scans reduce the influence of soft errors such as may be caused by random noise, and thus increase the likelihood that flaws will be accurately detected and decrease the likelihood that false errors will be reported~~. However, multiple flaw scans ~~are undesirable, as they increase manufacturing the time and required to manufacture a hard disk drive 100 and,~~

therefore, increase the cost of producing a hard disk drive ~~100~~ by decreasing manufacturing throughput.

There is, therefore, a need for a flaw scan that detects flaws and avoids false errors with high confidence ~~—Furthermore, with the limitations of conventional flaw scanning techniques, it is difficult to detect with high confidence the flaws that should be~~
5 ~~captured while rejecting false errors within fewer a single passes.~~ Therefore, conventional flaw scanning techniques require multiple passes if they are to detect flaws with high confidence.

~~—For the above stated reasons, it would be desirable to provide a method and an~~
10 ~~apparatus capable of detecting flaws in a disk 108 with improved confidence. In addition, it would be advantageous to provide a method and an apparatus for detecting flaws on a disk 108 that can provide the required confidence in as few a number of scans as possible, and preferably in one pass. It would also be desirable to provide a method and an apparatus that can detect flaws on a disk with a high degree of statistical~~
15 ~~confidence, without the detection of false errors. Furthermore, it would be desirable to provide both a method and an apparatus for the detection of flaws on a disk drive 100 that are reliable in operation and is that are inexpensive to implement.~~

SUMMARY OF THE INVENTION

20 In accordance with the present invention, both a method and an apparatus for detecting flaws in disks of a hard disk drive are provided. The present invention detects generally allows flaws in storage media in a disk to be detected with a much higher degree of statistical confidence and thus, thus allowing fewer passes than with

conventional flaw scan techniques using existing devices such as a PRML channel-

Furthermore, the present invention produces an overall media qualification that is at least as good or better than the conventional method and does so in less time.

5 In an accordance with one embodiment of the present invention, detecting flaws
in a disk drive includes sampling a read signal provided by reading a data pattern from a
disk to obtain samples, obtaining significant samples from the samples, deriving a value
from the significant samples, and reporting a flaw if a comparison between the derived
value and a threshold value is unacceptable.

10 In another embodiment, the data pattern is an iT pattern that includes a magnetic
transition every i^{th} bit cell on a track in which it is written.

In another embodiment, the significant samples are taken at times corresponding
to expected peak and near peak values in the read signal, which in turn correspond to
magnetic transitions in the data pattern, and the significant samples each have an
amplitude greater than 50% of an amplitude of an isolated pulse in the read signal
15 and an
apparatus are provided in which information written to a disk of a hard disk drive is read
from the disk. The information may be written on the disk in an iT data pattern where i is
an integer number. A read signal derived from the information written to the disk is
periodically sampled. All or a subset of the samples derived from a discrete portion of
the disk are then used to derive a value. The derived value is then compared to a
20 threshold value. If the derived value is less than the threshold value, a flag is generated
to signal the detection of a flaw.

In accordance with a further embodiment of the present invention, the information
written to the disk is written in an iT pattern so as to produce a periodic signal in the

transducer head of the disk drive. This periodic signal may in turn encode a repeated pattern of data.

In accordance with still another embodiment of the present invention, the significant samples are obtained by filtering the samples data is written on the disk such
5 that the magnetic polarity of a bit cell signals a 1 or alternatively a 0. A test pattern of data is then written on the disk. A method or apparatus in accordance with the present invention may then consider those samples of the signal taken at times corresponding to peaks or near peak samples in the readback signal. The samples under consideration may be used to derive a value for comparison with a threshold value. Based upon the
10 comparison with the threshold value, a signal may or may not be generated to indicate the presence of a defect.

—According to one embodiment of the present invention, the derived value for comparison with a threshold amount is obtained by passing a selected number of samples using through a digital band pass filter. For example, This has the result of suppressing
15 the effects of noise on the sampled signal. Accordingly, in an embodiment in which the test the data pattern is written in a 2T data pattern; and the the derived value is obtained by passing n samples through a digital filter has a that in delay operator notation of 1 —
$$D^2 + D^4 \text{ — } D^6 \dots \pm D^{2n};$$
-where n is equal to the number of total number of samples under consideration within the window. As According to still another example,
20 embodiment in which the test data pattern is written in a 3T data pattern and, the derived value is obtained by passing n samples through a digital filter has a that in delay operator notation of is given by the expression
$$1 + D \text{ — } D^3 \text{ — } D^4 + D^6 + D^7 \dots$$
[—/+ Dⁿ⁻¹ —/+ Dⁿ].

~~————In accordance with yet another embodiment of the present invention, the derived value is a sum, an average or an integration of the magnitudes of the significant samples, or of difference values between an optimal value and the magnitudes of the significant samples a value for comparison with a threshold value is derived by~~

5 ~~calculating a sum of the absolute value of a selected number of the samples under consideration. According to another embodiment of the present invention, the absolute value of all or a selected plurality of the n samples under consideration are integrated. According to still another embodiment of the present invention, the value for comparison with a threshold amount is derived by summing together the magnitudes of m of the n~~
10 ~~samples under consideration and dividing by m to obtain an average sampled value, where m is an integer number.~~

~~————In accordance with a further embodiment of the present invention, an optimal sample level is subtracted from the absolute value of each of m samples to obtain an error value. The error values obtained for each of the m samples may then be summed,~~
15 ~~averaged or integrated. The result may then be compared to a threshold value and a defect indicated if the comparison is not favorable.~~

~~In accordance with still another embodiment, the comparison between the derived value and the threshold value is unacceptable if the derived value of the apparatus of the present invention, the samples are fed into a Finite Impulse Response filter (FIR) or~~
20 ~~continuous shift register of samples that are summed into a value on every i^{th} clock cycle from all or a selected plurality of the n samples under consideration. In addition, a memory register for storing a threshold value is provided. The derived summation or value from the FIR filter is compared to the threshold value register at the input of a~~

comparator. ~~The comparator generates a signal indicating a detected defect if the sum of the samples is less than the threshold value~~amount.

Further~~Additional~~ advantages of the present invention will become readily apparent from the following discussion, particularly when taken together with the
5 accompanying drawings.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a ~~diagrammatic representation of a top view of a conventional computer disk drive, with the cover removed;~~

10 Fig. 2 is a diagrammatic representation of a ~~magnetic storage disk;~~

Fig. 3 is a ~~functional flow chart diagram of the operation of a conventional system for detecting flaw scan on a disk;~~

Fig. 4A illustrates a data pattern written to a track ~~depicts magnetic polarizations in a cross-section of track contained on the~~ disk;

15 Fig. 4B illustrates magnetic transitions in the data pattern ~~depicts the magnetization of the cross-section of track illustrated in Fig. 4A;~~

Fig. 4C illustrates a read signal provided by reading the data pattern ~~an example voltage potential produced in the channel as a result of the pattern of magnetization depicted in Fig. 4A~~B;

20 Fig. 5 illustrates depicts a read ~~typical signal produced in a channel in response to passing the transducer head through a pattern of magnetic flux, in which the signal is influenced by the presence of intersymbol interference and a flaw;~~

Fig. 6 is a functional-flow chart of a diagram illustrating the operation of a system
for the detection of flaw_scan on a disk in accordance with the present invention; and

Fig. 7 illustrates a functional is a functional block diagram of hardware diagram
configured to implement a flaw scan in accordance with an embodiment of the present
5 invention.

DETAILED DESCRIPTION

With reference now to Fig. 4A, illustrates a data pattern written to the portion of a
track 132 along a cross-sectional portion of the track 132. The data pattern is written to
10 to which a repeated pattern of data has been written is schematically illustrated. The
section of track 132 illustrated in Fig. 4A can be seen to contain a number of bit cells
400a-400li. The arrows An arrow in the a-bit cells 400 indicates the magnetic polarity of
the bit cells 400. In bit cells 400a, 400b, 400e, 400f, 400i and 400j the magnetic polarity
in a first direction encodes a digital 1, and in bit cells 400c, 400d, 400g, 400h, 400k and
15 400l the magnetic polarity in a second direction encodes a digital 0. Thus, the data
pattern is a 2T data pattern and the digital characters If the magnetization of a bit cell in a
given direction is a 1 and the magnetization of a bit cell in the opposite direction is a 0, it
can be seen that the pattern of data produced by the pattern of magnetic transitions
illustrated in Fig. 4A is such that each character is alternately repeated for two bit cells
20 400. That is, the pattern of bits is 1,1,0,0,1,1,0,0 . . . as shown in Fig. 4A. This is known
as a 2T data pattern. In a 3T pattern, the pattern of bits is 1,1,1,0,0,0,1,1,1,0,0,0 In
general, the pattern of magnetic transitions illustrated in Fig. 4A may be described as an
iT data pattern, where i is an integer number.

With reference now to Fig. 4B, illustrates the magnetic transitions in the data
pattern a pattern of magnetic polarities corresponding to the pattern of magnetization
shown in Fig. 4A is illustrated. The bit cells 400 as magnetized by the data pattern As
shown in Fig. 4B, a pattern of magnetic polarities written to a track 132 effectively forms
5 a series of magnets 404 in the track 132. As can be appreciated, at the boundaries
between of the magnets 404, which correspond to the boundaries between the bit cells 400
containing opposite magnetic polarities. Thus, the magnetic transitions occur at the
boundaries between the bit cells 400b and 400c, 400d and 400e, 400f and 400g, 400h and
400i, and 400j and 400k. Furthermore, the lines of magnetic flux produced by the
10 magnets 404 is will be normal to the surface of the disk 108 at in the vicinity of the
boundaries and disk surface. In addition, the direction of the magnetic flux will be
substantially parallel to the surface of the disk 108 in areas away from the
boundaries. magnetic transitions in a longitudinal recording scheme. It should be noted
that the present invention is equally applicable to a perpendicular recording scheme. As
15 can also be appreciated, a transducer head 124 following a track 132 in close proximity to
the surface of the disk 108 can sense these changes in magnetic flux. As will be
understood by those of ordinary skill in the art, the changes in magnetic flux can be used
to produce a voltage or a change in resistance in a read element of the transducer head
124.

20 With reference now to Fig. 4C, illustrates a read signal waveform 408 provided
duced by in the channel 140 as the transducer head 124 as it passes through the magnetic
flux produced by the bit cells 400 and reads the data pattern from the disk
108 magnetization illustrated in Fig. 4A is depicted. The read signal 408 includes peaks

412 that correspond to the magnetic transitions and zero-crossings 416 midway between
the magnetic transitions. In general, the waveform 408 is expressed in the channel 140 as
a voltage or current signal.

By comparing Figs. 4A and 4C, it can be appreciated that peaks 412 in the
5 waveform 412 occur periodically. In particular, the peaks 412 correspond to magnetic
transitions written on the track 132. Because the amplitude at the zero crossings is
expected to be zero, such samples can be discarded. Discarding samples taken at times
corresponding to zero crossings improves the signal to noise ratio of an asynchronously
sampled signal, as the amplitude of the retained samples, which are expected to be peak
10 (or near peak) values, will be relatively large compared to the near zero samples, where
noise can greatly affect the relative sample amplitude.

The waveform 408 illustrated in Fig. 4C is affected by intersymbol interference.
That is, the illustrated waveform 408 is produced in connection with magnetic transitions
within the track 132 that are close enough together to cause the magnetic polarizations
15 written to the track 132 to affect the flux patterns produced by adjacent magnetic
polarizations. In general, as bit cells 400 become smaller, the pattern of magnetic flux
produced by magnetic transitions between the bit cells 400 become increasingly subject
to intersymbol interference (*i.e.*, as the user bit density becomes greater than 1.0).

Fig. 5 illustrates a read signal 500 influenced by intersymbol interference and a
20 flawA waveform 500, such as may be produced by the pattern of magnetic transitions
illustrated in Fig. 4A in a typical disk drive 100 in which the waveform 500 is affected by
intersymbol interference, is illustrated in Fig. 5. It can be appreciated that any
disturbance of the system, such as may be produced by noise, can greatly affect how the

~~waveform 500 is interpreted. Therefore, it is useful to increase the signal to noise ratio in such systems. As mentioned above, it is possible to increase the signal to noise ratio by using a 2T or greater pattern and by considering only those samples taken at times that a significant non-zero value (i.e. a peak or near peak value) is expected. The read signal~~

5 500 has a irregular waveform shape due to intersymbol interference. The read signal 500 includes peak 504 with optimal amplitude and

~~—— With continued reference to Fig. 5, it can be seen that five peaks 508a-508e with attenuated of the example waveform 500 have an amplitude that is diminished amplitude relative to the other peaks. Since tTheis attenuated diminished amplitude is significantly~~
10 diminished and occurs in may be the result of noise and/or a flaw in the disk 108.

~~However, it is unlikely that noise would cause five peaks in a row, it is unlikely that the attenuated amplitude is due to noise to be as deeply attenuated as those illustrated in Fig. 5. Instead~~Therefore, the attenuated amplitude attenuation is probably likely due to a flaw in the disk 108.

15 Conventional flaw scan may not detect this flaw—Accordingly, this attenuation is preferably detected by the channel 140 and reported to the controller 136 of the disk drive 100. However, such a defect may not be detected using conventional flaw detection schemes. CFor example, onventional flaw scan may require a greater number of consecutive attenuated peaks than five in a row may be required. CIn addition,
20 eonventional flaw scan is also intechniques for detecting flaws in a disk drive 100 are less sensitive to slight variations in amplitude loss, and .—Therefore, if the read signal waveform 500 illustrated in Fig. 5 contains a particularly deeply diminished peak flaw, such as illustrated asby first alternate peak 512, then a conventional flaw scan does

detection system will not take this into consideration the relatively large attenuation of the first alternate peak 512. Furthermore, a conventional flaw scan detection system may fail to signal the detection of a flaw if even one of the peaks 508, illustrated as alternate peak 516, in a series has an amplitude greater than the threshold value. Therefore, even in connection with a flaw detection system that would detect a flaw after a series of five diminished peaks 508a-e, if even one of the peaks has an amplitude greater than the threshold (e.g., second alternate peak 516), no flaw will be detected.

Fig. 6 is a flow chart of a flaw scan in accordance with the present invention. The transducer head 124 writes a data pattern to the data fields 204 (step 600) and then reads the data pattern from the data fields 204 to obtain n-1 samples (step 604) and then a next sample (the nth sample) (step 608).

The channel 140 filters the n samples using a digital band pass filter to obtain m significant samples from the n samples (step 612). The significant samples are taken (sampled) at times corresponding to the expected peak and near peak values in the read signal, which in turn correspond to the magnetic transitions in the data pattern read from the disk 108. The significant samples each have an amplitude greater than 50% of an amplitude of an isolated pulse in the read signal. Furthermore, the significant samples each have an amplitude greater than the other samples of the n samples. Thus, the filtering passes the significant samples with the largest amplitudes and discards the other samples. For instance, the filtering passes the significant samples taken at or near the peaks 412 and discards the samples taken at or near the zero-crossings 416.

For example, the data pattern is a 2T data pattern and the filter has a delay operator notation of $1 - D^2 + D^4 - D^6 \dots \pm D^{2n}$. As another example, the data pattern is a

3T data pattern and the filter has a delay operator notation of $1 + D - D^3 - D^4 + D^6 + D^7$.
... $[-/+ D^{n-1} -/+ D^n]$. In either case, the filtering inverts various samples so that the
significant samples have the same sign, and the significant samples are determined in
accordance with the data pattern and the partial response of the channel 140.

5 Advantageously, the filtering increases the signal-to-noise ratio by retaining only
the peak and near peak samples taken at times corresponding to the magnetic transitions
in the data pattern and discarding the other samples where noise can greatly affect the
signal amplitude. In particular, the filtering reduces the noise bandwidth by the square
root of $1/m$ where m is the number of the significant samples that are considered. As a
10 result, the channel 140 more accurately distinguishes flaws from noise.~~According to the~~
~~present invention, a waveform or signal (e.g., 412 or 500) produced by reading a pattern~~
~~of magnetic transitions is analyzed using a peak filtering window method. In general, a~~
~~selected set, or window, of samples is considered as a group in order to reduce the effects~~
~~of noise on the sampled signal.~~

15 The channel 140 selects a predetermined number of the previous significant
samples using a moving window on a first-in first-out (FIFO) basis (step 616) and derives
a value based on the selected significant samples (step 620). As examples, the derived
value is a sum, an average or an integration of the magnitudes of the significant samples,
or a sum, an average or an integration of difference values between an optimal value and
20 the magnitudes of the significant samples.

The channel 140 determines whether the derived value is less than a threshold
value (step 624). If not, then the channel 140 returns to step 608 to take a next sample.

Otherwise, the channel 140 reports a flaw to the controller 136 (step 628) and returns to step 608 to take a next sample.

~~For~~ In addition, the signal to noise ratio may be increased by considering only samples taken at times corresponding to magnetic transitions (i.e. by considering only peak or near peak samples). According to one embodiment of the present invention, the magnitude or absolute value of the amplitude of m of the previous n samples under consideration are summed, or are integrated or averaged over m , to provide a value that can be compared to a selected threshold amount. Accordingly, the present invention effectively passes the samples under consideration through a band pass filter, reducing the noise bandwidth present in the signal by where m is the number of samples in the window that are considered. By reducing the noise in the signal, the signal to noise ratio is improved, and flaws in the disk 108 can be more accurately distinguished from noise. Accordingly, a more efficient technique for detecting flaws in a disk drive 100 is provided.

—— In accordance with another embodiment of the present invention, the previous n samples of a signal produced by an iT data pattern are passed through a filter that considers only those samples corresponding to expected peaks or near peaks in the signal derived from the data encoded on the disk 108. In delay operator notation, the filter in accordance with this embodiment of the present invention for use in connection with a 2T data pattern is given by the expression $1 - D^2 + D^4 - D^6 \dots \pm D^{2+n}$, where n is equal to the total number of samples within the window. A filter in accordance with the present invention for use in connection with a 3T data pattern is given by the expression $1 + D - D^3 - D^4 + D^6 + D^7 \dots [/+ D^{n-1} /+ D^n]$. As can be appreciated, these filters consider

every sample corresponding to an expected peak or near peak sample in the signal derived from the written pattern, and changes the sign of the sampled values so that they are all the same. As can also be appreciated, a digital filter that passes only the most significant samples (*i.e.* samples that are expected to contain peak or near peak values) and discards the other samples can be provided for any pattern of data. A significant sample may be understood to be any signal having an amplitude that is greater than 50% of the amplitude of an isolated pulse. As will be appreciated by those of ordinary skill in the art, the partial response of the channel's detector and the data pattern chosen for flaw scan detection will affect what samples are considered significant. The sum resulting from the application of such a filter may then be compared to an acceptable or a threshold value.

As an example, if the encoded data has a 2T pattern, and m is equal to 5, the digital filter applied to the samples is given by the expression $1 - D^2 + D^4 - D^6 + D^8$. As a further example, if the samples can be quantized into integer values ranging from -30 to +30, and the optimal sample value is 16 defined by the partial response of the channel, the sum obtained by the digital filter should be $m \times 16 = 80$ where m is equal to 5.

According to another embodiment of the present invention, the magnitude or absolute value of each of the previous significant samples are summed together. The resulting positive value is then compared to a selected acceptable or threshold value. The controller 136 is signaled that a defect has been detected if the sum of the samples are found to be unacceptable (*e.g.*, not greater than the acceptable value).

According to still another embodiment of the present invention, the sum resulting from the addition of the magnitude or absolute value of each of the previous m significant

samples is divided by m to obtain an average sampled value. The resulting average value, which will again be a positive number, may then be compared to a selected acceptable or threshold value. A defect is signaled to the controller 136 if the average value is unacceptable (e.g., is less than the threshold amount).

5 According to still another embodiment of the present invention, the magnitude or absolute value of each of the previous m significant samples are integrated. The result of the integration is then compared to a selected acceptable or threshold value. If the integrated value is less than the selected acceptable value, a defect is indicated and the controller 136 is signaled accordingly. example, the data pattern is a 2T data pattern,
10 m is equal to 5, the filter has a delay operator notation of $1 - D^2 + D^4 - D^6 + D^8$, the samples are quantized into integer values ranging from -30 to $+30$, the partial response of the channel 140 defines the optimal peak amplitude as 16, the derived value is a sum of the significant samples and the sum is $5 \times 16 = 80$.

15 With reference now to Fig. 6, the operation of an embodiment of the present invention is illustrated. Initially, at step 600, a pattern of magnetizations is written to the disk 108. For example, in a magnetic digital system, the magnetization of a bit cell in a first direction may encode a value 1, and the magnetization of a bit cell in second direction may encode a value 0.

20 A repeated pattern of magnetization and therefore of encoded data, is preferably written to the disk 108 to simplify the determination of whether a bit cell produces a signal of the expected amplitude in the channel 140. In a preferred embodiment of the present invention, the pattern of encoded information is written to all areas of a disk 108

that can be written to after the disk drive 100 has been assembled, including the data sectors 204 and all writable areas of the servo sectors 208. As is also noted above, the data may be written in an iT pattern, where i is an integer number, in which the direction of magnetic polarization is alternated every i bit cells. Although improvements in the signal to noise ratio may generally be realized by increasing the value of i (at least until the effective channel bit density is equal to one) it should be appreciated that the present invention is operable with any pattern of encoded data, including a 1T pattern.

Next, samples are taken from the signal derived from the transitions written to the disk. The signal or waveform is sampled once for every bit cell 400 in the length of track 132 in the window or length of track under consideration. As noted above, only the significant samples, such as the peak samples, which may include near peak samples, are preferably considered in order to increase the signal to noise ratio in the flaw detection system. Therefore, at step 604 $m-1$ of the samples are considered. That is, of the previous n samples taken, the $m-1$ samples corresponding to significant samples in the waveform are considered. At step 608, a next, or m^{th} , peak sample is taken. The magnitudes or absolute value of the amplitudes of each of the m peak samples are summed (step 612). Alternatively, the sign of the samples may be changed so that they are all the same, and the sum of the samples may then be calculated. As still another alternative, a difference between an absolute value of each of said m values and an optional value may be calculated, and the resulting differences may be added to obtain a sum.

At step 616, the sum derived from the previous m peak samples is compared to the selected threshold value (i.e., the acceptable value). If the sum is determined to be

~~less than the selected threshold value, the sum is unacceptable, and the controller 136 is signaled that a defect has been detected (step 620). Alternatively, the sum may be found to be unacceptable and the controller 136 may be signaled that a defect has been detected if the sum is not greater than the threshold value. The system then returns to step 608.~~

5 ~~—— If the sum of the previous m peak samples is determined to be acceptable, the system returns to step 608, at which point a next peak sample is taken. Accordingly, it can be appreciated that a moving window that includes the previous m peak samples is considered when determining whether to signal the controller 136 that a flaw requiring attention has been detected. That is, the previous m samples are considered on a first in~~
10 ~~first out (FIFO) basis. In addition, it can be appreciated that the effects of noise are diminished by considering a value derived from peak and near peak samples. That is, according to the present invention, the sample signal is subject to a band pass filter.~~

~~—— Although in a preferred embodiment, the m samples considered correspond to the m significant samples, it should be appreciated that, according to the present invention, m~~
15 ~~may be equal to n, the total number of samples within the window.~~

~~The selection of a threshold value will depends on the partial response of the disk drive's 100-channel 140. For example, where the read signal derived from the disk 108 may be is quantized into integer values ranging from -30 to +30, and the magnitude of the optimal peak signal amplitude is 16, a threshold value amount of less than 16 is~~
20 ~~would be selected for comparison with an average of the absolute value of each of the previous m significant samples. Likewise, a threshold value of less than $m \times 16$ is selected for comparison with a~~ Of course, where the sum or integrated value of the absolute values of the previous m significant samples are to be compared to a threshold

amount, the threshold value would be less than $m \times 16$. ~~A~~ According to one embodiment of the present invention, the threshold value to detect defects is about 50-90% of the accumulated value is suitable of the expected samples. The threshold value also depends on the size of the defects to be detected.

5 ~~With reference now to Fig. 7~~ illustrates, a functional hardware diagram to implement a flaw scan components in accordance with ~~an embodiment of the present invention are depicted in functional block diagram form~~. A shift register 700 receives the significant samples from the filter (not shown) on a FIFO basis and is provided for temporarily stores the significant samples as the ing absolute values of their

10 magnitudes ~~the previous m significant or peak values from the window of n samples~~. The ~~The provided samples may be stored as absolute values (i.e., they may represent a magnitude of the value, without reference to whether the value is positive or negative).~~ ~~The values stored in the shift register 700 are continually feeds the significant samples~~

15 ~~into a summing block 704. The summing block 704 calculates the derived value as a sum of the significant samples and the derived value (sum) output from the summing block 704~~ is continually clocked to a comparator 708. A memory 712 provides the threshold value ~~Also provided to the comparator 708 is a threshold value stored in memory 712.~~

20 ~~The output of the comparator 708 compares the sum with the threshold value and sends a flaw detect may comprise a signal provided to the controller 136 to indicate the detection of a defect if the sum of the previous m peak samples is unacceptable. For example, according to one embodiment of the present invention, the controller 136 is signaled if the sum of the m samples is less than the threshold value. In this manner, the shift~~

register 700, the summing block 704 and the comparator 708 implement steps 616, 620, and 624 and 628, respectively.

~~Although the hardware components that may comprise the present invention are illustrated as discrete components in Fig. 7, the example of Fig. 7 is intended to be non-limiting. Furthermore, it should be appreciated that the present invention may be implemented as software code running on a microprocessor. In addition, when used in connection with a hard disk drive 100, the present invention may be implemented as firmware code running in the controller 136 and/or channel 140 of the hard disk drive 100.~~

10 ~~Furthermore, although the present invention has been described in connection with the examples concerning magnetic disk drive 100s, the present invention is not so limited. In particular, the present invention may be applied to in connection with any storage memory device such as . For instance, the present invention may be applied in connection with optical, tape and/or three-dimensional storage devices. Similarly, the~~
15 present invention may be implemented in the disk drive 100 as software code running on a microprocessor or as firmware code running in the controller 136 and/or channel 140.
Likewise, although the present invention has been described in connection with a longitudinal recording disk 108, the present invention is equally applicable to a perpendicular recording disk. And although the signal-to-noise ratio can be increased by
20 increasing the period of an iT data pattern (at least until the effective channel bit density is one), the present invention is applicable to any data pattern including a 1T data pattern.

The foregoing discussion of the invention has been presented for purposes of illustration and description. Further, the description is not intended to limit the invention to the form disclosed herein. Consequently, variations and modifications commensurate with the above teachings, within the skill and knowledge of the relevant art, are within the scope of the present invention. The embodiments ~~described hereinabove~~ are further intended to explain the best mode presently known of practicing the invention and to enable others skilled in the art to utilize the invention in such or in other embodiments and with various modifications required by their particular application or use of the invention. It is intended that the appended claims ~~be construed to include the alternative~~ embodiments to the extent permitted by the prior art.

ABSTRACT

~~DA method and apparatus for detecting flaws in a disk included as part of a hard disk drive includes sampling a read signal provided by reading a data pattern from a disk to obtain samples, obtaining significant samples from the samples, deriving a value from the significant samples, and reporting a flaw if a comparison between the derived value and a threshold value is unacceptable.~~ A signal derived from data encoded on the disk is sampled. All or a subset of the n samples derived from a discrete portion of the disk are then used to derive a value. For example, the previous m significant samples are used to derive a value. The value may be derived by calculating a sum of the previous m samples, by integrating the previous m samples, by calculating an average of the previous m samples or by passing the previous m samples through a filter. The derived value is then compared to a threshold value. If the derived value is less than the threshold value, the controller of the disk drive is signaled to indicate that a flaw has been detected. The present invention allows flaws on disks or other storage media to be detected with improved accuracy.